

## $\beta$ -expansion's Attractors Observed in A/D converters

Tohru Kohda<sup>1</sup>, Yoshihiko Horio<sup>2</sup>, and Kazuyuki Aihara<sup>3</sup>

<sup>1</sup>Department of Informatics, Kyushu University, Motoooka 744, Nisi-ku, Fukuoka, 819-0395, Japan,  
kohda@inf.kyushu-u.ac.jp

<sup>2</sup>Department of Electrical and Electronic Eng., Tokyo Denki University, Tokyo, 101-8457, Japan,  
horio@eee.dendai.ac.jp

<sup>3</sup>Institute of Industrial Science, The University of Tokyo, 4-6-1, Komaba Meguro-ku, Tokyo, 153-8505, Japan,  
aihara@sat.t.u-tokyo.ac.jp

A new class of analog-to-digital (A/D) and digital-to-analog (D/A) converters using a flaky quantiser, called the  $\beta$ -encoder, as shown in Fig. 1 [1, 2, 3] has been shown to have exponential bit rate accuracy while possessing a self-correction property for fluctuations of the amplifier factor  $\beta$  and the quantiser threshold  $\nu$ . Motivated by the close relationships [4, 5, 6] between  $\beta$ -transformations and  $\beta$ -expansion, we have recently observed [9, 10] that (1) such a flaky quantiser is exactly realized by the “multi-valued Rényi-Parry map”, defined here so that probabilistic behavior in the “flaky region” is completely explained using dynamical systems theory; (2) a sample  $x$  is always confined to a subinterval of the contracted interval while the successive approximation of  $x$  is stably performed using  $\beta$ -expansion even if  $\nu$  may vary at each iteration (i.e. a small real-valued quantity, approximately proportional to the quantisation error, does not necessarily converge to any fixed value, e.g., 0 but may oscillate without diverging). Such a phenomenon is precisely the kind of “chaos”; (3) such a subinterval enables us to obtain the decoded sample easily, as it is equal to the midpoint of the subinterval and to prove two classic  $\beta$ -expansions, known as the greedy and lazy expansions [7, 8] are perfectly symmetrical in terms of their quantisation errors. The subinterval further suggests that  $\nu$  should be set to around the midpoint of its associated greedy and lazy values. A switched-capacitor (SC) circuit technique [11, 12] has been proposed for implementing A/D converter circuit based on several types of  $\beta$ -encoders and SPICE simulations have been given to verify the validity of these circuits against deviations and mismatches of circuit parameters. Our review

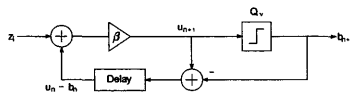


Figure 1: A typical  $\beta$ -encoder with its input  $z_0 = y \in [0, 1)$ ,  $z_i = 0, i > 1$  and output  $(b_{i,\beta})_{i \geq 1}$ .

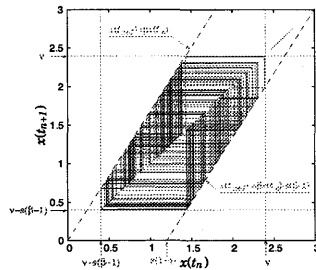


Figure 2: The scale-adjusted  $\beta$ -map:  $S_{\beta,\nu,s}(x)$  and its eventually onto map in which an attractor can be observed.

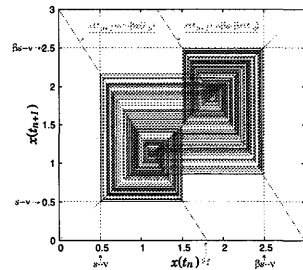


Figure 3: The scale-adjusted negative  $\beta$ -map:  $R_{\beta,\nu,s}(x)$  and its eventually onto map in which an attractor can be observed.

is twofold. First, the  $\beta$ -encoder leads us to naturally define the “multi-valued Rényi-Parry map” [4, 5] with its eventually onto map, as it is identical to the Parry’s  $(\beta, \alpha)$ -map [6]. Second, chaos, called “ $\beta$ -expansion’s attractors” can be observed on the onto-map. Two types of  $\beta$ -expansion’s attractors are as follows:

1. **Scale-Adjusted  $\beta$ -Map**[9, 11]: Daubechies et al. [1, 2] introduced a “flaky” version of an imperfect quantiser, defined as

$$Q_{\Delta_\beta}^f(z) = \begin{cases} 0, & \text{if } z \leq \nu_0, \\ 1, & \text{if } z \geq \nu_1, \\ 0 \text{ or } 1, & \text{if } z \in \Delta_\beta = [\nu_0, \nu_1], \nu_0 < \nu_1, \end{cases} \quad (1)$$

which is a  $\nu$ -varying model of a quantiser  $Q_\nu(z) = \begin{cases} 0, & \text{if } z \leq \nu, \\ 1, & \text{if } z \geq \nu, \end{cases} \nu \in [\nu_0, \nu_1], \nu_0 < \nu_1$ . We obtain:

**Lemma 1**[9]: Let  $S_{\beta,\nu,s}(x)$  be the scale-adjusted map with a scale  $s$ , defined by

$$S_{\beta,\nu,s}(x) = \beta x - s(\beta - 1)Q_{\gamma\nu}(x) = \begin{cases} \beta x, & x \in [0, \gamma\nu), \\ \beta x - s(\beta - 1), & x \in [\gamma\nu, s), \end{cases} \nu \in [s(\beta - 1), s), s > 0 \quad (2)$$

which is referred to as the “multi-valued Rényi-Parry map” on the flaky region  $\Delta_\beta = [s(\beta - 1), s]$  and has its eventually onto Parry’s  $(\beta, \alpha)$ -map [6] with the subinterval  $[\nu - s(\beta - 1), \nu]$  as shown in Fig.2. This map realises the flaky quantiser  $Q_{[s(1-\gamma), s\gamma]}^f(\cdot)$ . Let  $b_{i, \beta, \nu_1^i, s}$  be its associated bit sequence for the threshold sequence  $\nu_1^L = \nu_1 \nu_2 \cdots \nu_L$ , defined by

$$b_{i, \beta, \nu_1^i, s} = Q_{\gamma\nu}(S_{\beta, \nu_1^{i-1}, s}^{i-1}(x)) = \begin{cases} 0, & S_{\beta, \nu_1^{i-1}, s}^{i-1}(x) \in [0, \gamma\nu], \\ 1, & S_{\beta, \nu_1^{i-1}, s}^{i-1}(x) \in [\gamma\nu, s]. \end{cases} \quad (3)$$

Then we get  $x = s(\beta - 1) \sum_{i=1}^L b_{i, \beta, \nu_1^i, s} \gamma^i + \gamma^L S_{\beta, \nu_1^L, s}^L(x)$  and its decoded value

$$\hat{x}_{L, \beta, \nu_1^L, s} = s(\beta - 1) \sum_{i=1}^L b_{i, \beta, \nu_1^i, s} \gamma^i + \frac{s\gamma^L}{2}.$$

## 2. Negative $\beta$ -Map[10, 12]: We get

**Lemma 2**[10]: Let  $R_{\beta, \nu, s}(x) : [0, s] \rightarrow [0, s]$ ,  $s > 0$  be the (scale-adjusted) negative  $\beta$ -map, defined by

$$R_{\beta, \nu, s}(x) = -\beta x + s[1 + (\beta - 1)Q_{\gamma\nu}(x)] = \begin{cases} s - \beta x, & x \in [0, \gamma\nu], \\ \beta s - \beta x, & x \in [\gamma\nu, s], \end{cases} \quad \nu \in [s(\beta - 1), s] \quad (4)$$

which is another “multi-valued Rényi-Parry map” on the flaky region  $\Delta_\beta = [s(\beta - 1), s]$  realising  $Q_{[s(1-\gamma), s\gamma]}^f(\cdot)$  and has its eventually onto Parry’s  $(\beta, \alpha)$ -map [6] with the subinterval  $[s - \nu, \beta s - \nu]$  as shown in Fig. 3. Let  $b_{i, \beta, \nu_1^i, s}$  be the associated bit sequence for the threshold sequence  $\nu_1^L$ , defined by

$$b_{i, \beta, \nu_1^i, s} = Q_{\gamma\nu}(R_{\beta, \nu_1^{i-1}, s}^{i-1}(x)) = \begin{cases} 0, & R_{\beta, \nu_1^{i-1}, s}^{i-1}(x) \in [0, \gamma\nu], \\ 1, & R_{\beta, \nu_1^{i-1}, s}^{i-1}(x) \in [\gamma\nu, s]. \end{cases} \quad (5)$$

Then we get  $x = (-\gamma)^L R_{\beta, \nu_1^L, s}^L(x) - s \sum_{i=1}^L f_{i, \beta, \nu_1^i, s} (-\gamma)^i$  and its decoded value

$\hat{x}_{L, \beta, \nu_1^L, s} = s\{(-\gamma)^L/2 - \sum_{i=1}^L f_{i, \beta, \nu_1^i, s} (-\gamma)^i\}$ . where  $f_{i, \beta, \nu_1^i, s} = 1 + b_{i, \beta, \nu_1^i, s}(\beta - 1)$ . Such a negative  $\beta$ -expansion defines a new A/D converter called a *negative  $\beta$ -encoder* which facilitates the implementation of stable analog circuits. Figures 2 [11] and 3 [12] show a typical  $\beta$ -expansion’s attractor of Eqs.(2) and (4), respectively.

## Acknowledgments

This research is supported by the Japan Society for the Promotion of Science (JSPS) through its “Funding Program for World-Leading Innovative R&D on Science and Technology (FIRST Program)”.

## References

- [1] I. Daubechies, R. DeVore, C. Güntürk, and V. Vaishampayan, “Beta expansions: a new approach to digitally corrected A/D conversion,” *IEEE ISCAS 2002*, **2**, 784-787, May. 2002.
- [2] I. Daubechies, R. DeVore, C. Güntürk, and V. Vaishampayan, “A/D conversion with imperfect quantizers,” *IEEE Transactions on Information Theory*, **52-3**, 874-885, Mar. 2006.
- [3] I. Daubechies, and Ö. Yilmaz, “Robust and practical analog-to-digital conversion with exponential precision,” *IEEE Transactions on Information Theory*, **52-8**, 3533-3545, Aug. 2006.
- [4] A. Rényi, “Representations for real numbers and their ergodic properties,” *Acta Mathematica Hungarica*, **8**, no.3-4, 477-493, Sep. 1957.
- [5] W. Parry, “On the  $\beta$ -expansions of real numbers,” *Acta Math. Acad. Sci. Hung.*, **11**, 401-416, 1960.
- [6] W. Parry, “Representations for real numbers,” *Acta Math. Acad. Sci. Hung.*, **15**, 95-105, 1964.
- [7] P. Erdős, and I. Joó, “On the expansion  $1 = \sum_{i=1}^{\infty} q^{-n_i}$ ,” *Periodica Mathematica Hungarica*, **23-1**, 25-28, Aug. 1991.
- [8] P. Erdős, I. Joó and V. Komornik, “Characterization of the unique expansions  $1 = \sum_{i=1}^{\infty} q^{-n_i}$  and related problems”, *Bull. Soc. Math. France*, **118**, 377-390, 1990.
- [9] Hironaka, S., Kohda, T. and Aihara, K., “Markov chain of binary sequences generated by A/D conversion using  $\beta$ -encoder”, *Proc. of 15th IEEE International Workshop on Nonlinear Dynamics of Electronic Systems*, 261-264, 2007.
- [10] T. Kohda, S. Hironaka, and K. Aihara, “Negative  $\beta$ -encoder,” arXiv:0808.2548v1 [cs.IT] 19 Aug 2008(arXiv:0808.2548v2 [cs.IT] 28 Jul 2009).
- [11] Y. Horio, K. Jin’no, T. Kohda and K. Aihara. “Circuit Implementation of an A/D Converter Based on the Scale- Adjusted  $\beta$ -Map Using a Discrete-Time Integrator,” *18th IEEE Workshop on Nonlinear Dynamics of Electronic Systems (NDES2010)*, 110-113, 2010.
- [12] Y. Horio, K. Jin’no, T. Kohda and K. Aihara. “Circuit Implementation of an A/D Converter Based on the Negative  $\beta$ -Map with a Discrete-Time Integrator,” *2010 International Symposium on Nonlinear Theory and its Applications (Nolta2010)*, 265-268, 2010.