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Fabrication and Characteristics of HTS Induction Motor by the Use of Bi-2223/Ag Squirrel-Cage Rotor

Taketsune Nakamura, Hironori Miyake, Yoshio Ogama, Gaku Morita, Itsuya Muta, and Tsutomu Hoshino

Abstract—HTS squirrel-cage induction motor was fabricated and tested in this study. Both of rotor bars and end rings were made of Bi-2223/Ag multifilamentary tapes in order to realize superconducting current loops, and the conventional (normal conducting) stator, 3-phase and 4-pole, was utilized. Rotating characteristics of the fabricated motor were tested for different input voltages at 60 Hz. The performances were also analyzed by means of the theoretical method based on the electrical equivalent circuit. It was shown that the minimum starting voltage was experimentally confirmed and agreed semi-quantitatively with the analysis result. The rotation at synchronous speed was realized by applying the load, at least, 1.5 Nm.

Index Terms—Bi-2223/Ag multifilamentary tape, induction motor, minimum starting voltage, persistent current mode, synchronous torque.

I. INTRODUCTION

INDUCTION motor is well known to be widely utilized in a capacity range from fractional horsepower to large capacity motors. For instance, all of the inverter fed motors for the electric trains are squirrel-cage type induction motors in Japan. On the other hand, large current will be induced in the end rings at 6 turns, i.e., its total critical current is about 540 A (= 25 A × 4).

On the other hand, large current will be induced in the end rings compared to the rotor bars, and then the tape that has large critical current, I_c (= 90 A at 77.3 K and self-field), is utilized for this purpose. Here, the HTS end rings are designed to be always in the superconducting state, i.e., the induced current will be less than I_c. That is, the end rings consist of the Bi-2223/Ag windings at 6 turns, i.e., its total critical current is about 540 A (= 90 A × 6).

II. FABRICATION AND TEST METHOD

A. Specifications of Bi-2223/Ag Tapes

Both of rotor bars and end rings are fabricated by using Bi-2223/Ag multifilamentary tapes. The specifications, however, are different between them (Table I). That is, narrow width tape, i.e., 2.0 mm, is utilized for rotor bars because of the restriction of the space in the rotor slot. Four tapes are connected in parallel by using solder, and this bundle is used for a rotor bar, i.e., its total critical current is 100 A.

B. Characterization of Bi-2223/Ag Tape for Rotor Bars

The performance of the induction motor is dominated by the characteristics of Bi-2223/Ag tapes utilized for the rotor bars. Therefore, electric field, E, vs. current density, J, (E ∝ J) curve for such tape is firstly characterized. It should be noted that the HTS rotor bars are placed in the slots of the rotor, which are made of laminated silicon steel cores. Then, almost all magnetic fluxes pass through this iron rotor, and the HTS rotor bars experience only self-field. In other words, we don’t need to take account of the magnetic field dependency of current transport properties in Bi-2223/Ag tape, i.e., self-field application. Fig. 1 shows the double-logarithmic plot of E ∝ J curve of the tape for...
rotor bars, measured at 77.3 K and self-field. In this figure, the empty circles denote measured data, and the solid curve shows the theoretical one (Weibull function), which is expressed as follows [4], [5]

\[ E(J) = \rho_{\ell\ell} J_0 \int_0^{J \sim J_{\text{cm}}} \left[ 1 - \exp \left( -\frac{J}{J_{\text{cm}}} \right)^n \right] \, dJ, \quad J > J_{\text{cm}} \quad (1) \]

\[ 0, \quad J \leq J_{\text{cm}} \quad (2) \]

where \( J_{\text{cm}} = 6.27 \times 10^7 \, \text{A/m}^2 \) and \( J_0 = 5.19 \times 10^8 \, \text{A/m}^2 \), respectively, denote the minimum and the half value of width of the local critical current density distribution. The parameter \( \rho_{\ell\ell} = 9.57 \times 10^{-8} \, \text{[\Omega m]} \) means the resistivity for uniform flux flow, and \( n = 5.72 \) the shape of the critical current distribution. Further, the effect of the current sharing to the silver sheath is also considered, and the total current, \( J_{\text{t}} \), is calculated by using the following relation

\[ J_{\text{t}} = \frac{1}{c+1} J + \frac{c}{c+1} \frac{E(J)}{\rho_{\text{Ag}}} \quad (3) \]

where, the parameter \( c = 2.0 \) and \( \rho_{\text{Ag}} \), respectively, denote the silver ratio and the resistivity of silver sheath.

As can be seen in Fig. 1, the agreement between the experiment and the theory is excellent, and then (1)–(3) are introduced to the analysis of the motor performance based on the electrical equivalent circuit.

C. HTS Rotor

As is already mentioned, the HTS squirrel-cage windings are fabricated by using the laminated silicon steel rotor. In this study, the stator of the commercialized induction motor, 3-phase, 4-pole, 1.5 kW, are utilized. In other words, the secondary windings are only replaced by HTS, and then the simple structure of the conventional induction motor is maintained. Fig. 2 illustrates the schematic diagram of the cross section of a rotor slot. As can be seen, the superconducting Bi-2223 rotor bar is put at the outermost surface of the rotor, i.e., near the air-gap, in order to reduce the effect of the leakage magnetic flux [3]. Insulated copper bars are also installed in the slot in order to support as well as cool the rotor bars. Bi-2223 end rings are connected with the above mentioned rotor bars by using the solder. Fig. 3 shows the photographs of (a) rotor core and slots; and (b) completed HTS rotor, respectively.

D. Test Method

The fabricated motor is immersed in liquid nitrogen, and then the rotating characteristics are tested at no-load and light load conditions. Before the experiment, the lubricating oil of the roller bearings are removed by using acetone for avoidance of freezing. Then, the mechanical loss of the bearings are enlarged due to the friction. Rotating speed, \( N_r \), and torque, \( \tau \), are measured by means of torque/speed meter installed outside the cryostat (Fig. 4). The application of light load is also available by using the powder brake (its maximum value is 1.8 Nm). The driving frequency is fixed to be 60 Hz, and then the corresponding synchronous speed, \( N_{srs} \), is 1800 rpm.

III. ANALYSIS METHOD

Performance of the fabricated HTS induction motor is also investigated based upon the commonly utilized equivalent circuit as shown in Fig. 5. In these figures, \( V_1 \) means the input voltage (phase voltage). And \( F'(I_2) / s \) and \( V_2 \), denote the voltage of the secondary (HTS) windings at slip and synchronous operations,
respectively. Here, slip, \( s \), is defined as \( s = (N_p - N)/N_p \). The parameters \( g_1, l_1, r_1, \) and \( x_1 \), correspondingly, denote the exciting conductance, exciting susceptance, resistance of the windings, and the leakage reactance. Further, the subscript 1 (or 2) shows the primary (or secondary) side, and the prime ‘\( \prime \)’ shows that the secondary parameter is converted into the primary side.

At slip mode [Fig. 5(a)], the nonlinear resistance, \( r_1'(I_2') \), due to the flux flow is generated in the HTS rotor windings. In this case, the voltage, \( F'(I_2')/s = r_1'(I_2')I_2' \), is calculated by using (1) and (3) (for detailed explanation, see [3] because of the lack of space). On the other hand, when the above-mentioned resistance becomes zero by accelerating, i.e., the corresponding slip frequency reduces, the persistent current mode will be realized. In that case, the interlinked magnetic fluxes will be trapped in the windings, and then the voltage source, \( V_{2s} \), equivalently appears at the secondary windings, i.e., synchronous mode [Fig. 5(b), see (5) for \( V'_{2s} \)].

\[ F'(I_2') = \left| V_1 - \sqrt{r_1'(I_2')^2 - (x_1 + x_2')^2|I_2'|^2} \right| \]

Then, the voltage for the synchronous mode [Fig. 5(b)], \( |V'_{2s}| \), can be calculated by using the relation, \( |I_2'| = |I_e'| \), as follows

\[ |V_{2s}| = \lim_{I_2' \rightarrow I_e'} \frac{F'(I_2')}{s} \]

where, the parameter, \( |I_e'| \), is the critical current defined by the electric field criterion at 1 \( \mu \)V/cm. Finally, the minimum starting voltage (phase voltage), \( |V_{1s}| \), is expressed by using (4) and (5) with the condition, \( V_{2s}' = 0 \), as follows.

\[ |V_{1s}| = \sqrt{r_1^2 + (x_1 + x_2')^2|I_e'|^2} \]

Based on the experiment, the parameters, \( r_1, (x_1 + x_2') \) and \( |I_e'| \) are determined to be, 0.12 \( \Omega \), 5 \( \Omega \) and 4 A, respectively. And then, the corresponding minimum starting voltage, \( |V_{1s}| \), is 18.5 \( V_{rms} \).

On the other hand, this voltage is also confirmed by the experiment. That is, \( |V_{1s}| \) is defined by the voltage at which the HTS motor starts rotating, and estimated to be 22.7 \( V_{rms} \) (see Fig. 6). This value is semi-quantitatively consistent with that of the analysis result. The reasons for the difference between them are friction of the liquid nitrogen, mechanical loss of the roller bearing, etc.
B. Rotating Characteristics for Different Input Voltages

Fig. 6 shows the test results of relationship between the rotating speed, $N$, vs. input voltage, $V_1$. All data are obtained at steady state. As is already mentioned, the motor starts rotating at $V_1 = 22.7 \text{V}_{\text{rms}}$. Further, it can be seen that the motor is swiftly pulled in to be synchronous speed (1800 rpm) just after starting.

Fig. 7 also shows the experimental results of the rotating characteristics at light load application. In case of $V_1 = 50 \text{V}_{\text{rms}}$, the motor stops rotating by applying the load at 0.8 Nm. As the input voltage increases, however, the motor rotates at slip mode. Moreover, the synchronous speed is available even for the application of the load at around 1.4 Nm, when the input voltage is higher. This means that the synchronous torque exists in our fabricated HTS induction motor. This torque is impossible to be obtained for the conventional (normal conducting) induction motor, because the relative speed between the gap magnetic flux and the rotor is zero at synchronous speed. Namely, both of slip and synchronous operation is possible for the HTS induction motor.

C. Load Test Results

As is already mentioned, the light load can only be applied by means of the powder brake in this test system. Then, the load characteristics are obtained from the equivalent circuit (see Section III). Fig. 8 shows the calculated results of torque, $\tau$, vs. slip, $s$, curves. The curve of the conventional motor is also plotted for comparison (broken curve). As can be seen, the starting and accelerating characteristics of the HTS induction motor improve drastically compared with those of the conventional motor. That is, the maximum torque at 50 Nm is obtained, and this value is more than 3 times larger compared with the conventional one. This indicates that special structural techniques, such as double-cage or deep slots, utilized for the conventional motor is unnecessary for obtaining adequate starting torque. Furthermore, the synchronous torque, $\tau_s$, at about 10 Nm is also realized. This is because of the trapping of the magnetic flux in the HTS squirrel-cage windings [see Fig. 5(b)]. In other words, the HTS induction motor is really interesting to possess larger starting torque, accelerating torque and synchronous torque.

V. CONCLUSION

The HTS squirrel-cage type induction motor was fabricated and tested in this study. The secondary windings were only fabricated by using the Bi-2223/Ag tapes, and the commercialized stator was utilized. The performance of the motor was also estimated based on the electrical equivalent circuit taking account of the nonlinear $E – J$ curve of Bi-2223/Ag tape. It was shown theoretically and experimentally that there exists the minimum voltage for starting. The HTS induction motor also had large starting as well as accelerating torque compared with the conventional motor. Furthermore, the synchronous torque was available by trapping the magnetic flux in the squirrel-cage windings.

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